

end-to-end average distortion with the SNR, which was referred to as the distortion SNR exponent $a^*(\eta)$, for given spectral efficiency η . We first derived an upper bound on $a^*(\eta)$ based on an informed transmitter that has instantaneous knowledge of the channel capacity. Then the exponent achievable with a separation based scheme was computed. Finally, we proposed HDA source-channel coding schemes that outperform the separated exponent for all η . Remarkably, the HDA scheme for $\eta > 2 \min(M, N)$ was shown to achieve the optimal distortion exponent for general M and N . We also showed how to construct practical space-time coding schemes using diversity-multiplexing tradeoff optimal space-time codes and scalar quantization.

REFERENCES

- [1] E. Biglieri, J. Proakis, and S. Shamai (Shitz), "Fading channels: Information-theoretic and communications aspects," *IEEE Trans. Inf. Theory*, vol. 44, no. 6, pp. 2619–2692, Oct. 1998.
- [2] L. Zheng and D. N. C. Tse, "Diversity and multiplexing: A fundamental tradeoff in multiple-antenna channels," *IEEE Trans. Inf. Theory*, vol. 49, no. 5, pp. 1073–1096, May 2003.
- [3] J. N. Laneman, E. Martinian, G. W. Wornell, J. G. Apostolopoulos, and S. J. Wee, "Comparing application- and physical-layer approaches to diversity on wireless channels," in *Intl. Conf. Communications*, Anchorage, AK, May 2003, pp. 2678–2882.
- [4] J. N. Laneman, E. Martinian, G. W. Wornell, and J. G. Apostolopoulos, "Source Channel Diversity for Parallel Channels," *IEEE Trans. Inf. Theory*, vol. 51, no. 10, pp. 3518–3539, Oct. 2005.
- [5] T. Holliday and A. J. Goldsmith, "Joint source and channel coding for MIMO systems," in *Proc. 42nd Annu. Allerton Conf. Communications, Control, and Computing*, Monticello, IL, Oct. 2004.
- [6] D. Gunduz and E. Erkip, "Source and channel coding for quasi-static fading channels," in *Proc. 39th Asilomar Conf. Signals, Systems and Computes*, Monterey, CA, Oct./Nov. 2005, pp. 18–22.
- [7] D. Gunduz and E. Erkip, "Source and channel coding for cooperative relaying," in *Proc. Int. Workshop on Signal Processing Advances for Wireless Communications (SPAWC)*, New York City, Jun. 2005, pp. 950–954.
- [8] G. Caire and K. R. Narayanan, "On the SNR exponent of hybrid digital analog space time codes," in *Proc. 43rd Annu. Allerton Conf. Communications, Control and Computing*, Monticello, IL, Oct. 2005.
- [9] K. R. Narayanan and G. Caire, "Further results on the SNR exponent of hybrid digital analog space time codes," in *Proc. UCSD Workshop on Information Theory and Its Applications*, San Diego, CA, Jan. 2006.
- [10] B. Hochwald and K. Zeger, "Tradeoff between source and channel coding," *IEEE Trans. Inf. Theory*, vol. 43, no. 5, pp. 1412–1424, Sep. 1997.
- [11] U. Mittal and N. Phamdo, "Hybrid digital-analog (HDA) joint source-channel codes for broadcasting and robust communications," *IEEE Trans. Inf. Theory*, vol. 48, no. 5, pp. 1082–1102, May 2002.
- [12] Z. Reznic, M. Feder, and R. Zamir, "Distortion bounds for broadcasting with bandwidth expansion," in *Proc. Conv. Electrical and Electronics Engineers in Israel*, Dec. 2002, p. 19.
- [13] B. P. Dunn and J. N. Laneman, "Characterizing source-channel diversity approaches beyond the distortion exponent," in *Proc. 43rd Annu. Allerton Conf. Communications, Control, and Computing*, Monticello, IL, Sep. 2005.
- [14] D. Gunduz and E. Erkip, "Joint source-channel codes for MIMO block fading channels," *IEEE Trans. Inf. Theory*, submitted for publication.
- [15] D. Gunduz and E. Erkip, "Distortion exponents of MIMO fading channels," in *Proc. IEEE Information Theory Workshop*, Punta del Este, Uruguay, Jul. 2006, pp. 694–698.
- [16] M. Gastpar, B. Rimoldi, and M. Vetterli, "To code, or not to code: Lossy source-channel communication revisited," *IEEE Trans. Inf. Theory*, vol. 49, no. 5, pp. 1147–1158, May 2003.
- [17] J. Proakis and M. Salehi, *Fundamentals of Communication Systems*. Upper Saddle River, NJ: Prentice-Hall, 2004.
- [18] T. M. Cover and J. A. Thomas, *Elements of Information Theory*. New York: Wiley, 1991.
- [19] A. Dembo and O. Zeitouni, *Large Deviations Techniques and Applications*. Berlin/New York: Springer-Verlag, 1998.
- [20] T. Holliday and A. J. Goldsmith, "Optimizing end-to-end distortion in MIMO systems," in *Proc. IEEE Symp. Information Theory*, Adelaide, Australia, Sep. 2005, pp. 1671–1675.
- [21] G. Caire and D. Tuninetti, "The throughput of hybrid-ARQ protocols for the Gaussian collision channel," *IEEE Trans. Inf. Theory*, vol. 47, no. 5, pp. 1971–1988, Jul 2001.
- [22] H. El Gamal, G. Caire, and M. O. Damen, "The diversity-multiplexing-delay tradeoff in MIMO ARQ channels," in *Proc. IEEE Int. Symp. Information Theory*, Adelaide, Australia, Sep. 2005, pp. 1823–1826.
- [23] H. El Gamal, G. Caire, and M. Damen, "Lattice coding and decoding achieve the optimal diversity-vs-multiplexing tradeoff of MIMO channels," *IEEE Trans. Inf. Theory*, vol. 50, no. 6, pp. 968–985, Jun. 2004.
- [24] P. Elia, B. Sethuraman, and P. Kumar, "Perfect space-time codes with minimum and nonminimum delay for any number of antennas," in *Int. Conf. Wireless Networks, Communications and Mobile Computing*, Maui, HI, Jun 2005, vol. 1, pp. 722–727.
- [25] P. Elia, K. Kumar, S. Pawar, P. Kumar, and H.-F. Lu, "Explicit space-time codes that achieve the diversity-multiplexing gain tradeoff," in *Proc. IEEE Int. Symp. Information Theory*, Adelaide, Australia, Sep. 2005, pp. 896–900.
- [26] J.-C. Belfiore, G. Rekaya, and E. Viterbo, "The Golden code: A 2×2 full-rate space-time code with nonvanishing determinant," *IEEE Trans. Inf. Theory*, vol. 51, no. 4, pp. 1432–1436, Apr. 2005.
- [27] S. Tavildar and P. Viswanath, "Permutation codes: Achieving the diversity-multiplexing tradeoff," in *Proc. IEEE Int. Symp. Information Theory*, Chicago, IL, Jun./Jul. 2004, p. 98.
- [28] R. M. Gray, *Source Coding Theory*. Berlin/ew York: Springer-Verlag, 1989.

A Note on the Optimality of Variant-I Permutation Modulation Codes

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Abstract—In this correspondence, the optimality of variant-I permutation codes initially proposed by Slepian is shown in a simple way.

Index Terms—Group codes, permutation codes, permutation modulation.

I. INTRODUCTION

In [1], two variants of permutation modulation (PM) codes are introduced. A variant-I code C is defined by l integers $m_1 \leq m_2 \leq \dots \leq m_l$ satisfying $\sum_{i=1}^l m_i = n$ and the n -dimensional vector

$$\mathbf{x} = (x_1, \dots, x_1, x_2, \dots, x_2, \dots, x_l, \dots, x_l) \quad (1)$$

where for $i = 1, \dots, l$, x_i appears m_i times. The codebook is formed of all

$$M = \frac{n!}{m_1! m_2! \dots m_l!} \quad (2)$$

possible distinct permutations of \mathbf{x} . The rate of the code is $R = \log_2 M/n$ and its minimum squared Euclidean distance (MSED) is defined as

$$d_{\min}^2 = \min_{\mathbf{x} \neq \mathbf{x}' \in C} \|\mathbf{x} - \mathbf{x}'\|^2. \quad (3)$$

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Based on the results of [2], the design of an optimum PM variant-I code can be formulated as follows: “Given the set $\{m_1, m_2, \dots, m_l\}$, find the corresponding set of points $\{x_1, x_2, \dots, x_l\}$ such that $d_{\min}^2 = 2$ and $\|\mathbf{x}\|^2$ is minimized.”¹ In fact, [2] provides the solution to this problem but for the last step of the proof, it refers to [3]. In the following, we provide an alternative answer to this part.

It can be noted that the optimum solution derived in [4] is more restrictive as the values m_i 's need to satisfy $m_i = m_{l-i-1}$. As a result, many optimum codes found by our approach are not considered by that of [4]. On the other hand, it is straightforward to show that the solution to our optimization design problem for variant-II PM codes is equivalent to that of [4].

II. OPTIMUM VARIANT-I CODES

In this section, the optimality of variant-I PM codes is derived based on the following.

Lemma 2.1: Consider l integers $m_1 \leq m_2 \leq \dots \leq m_l$ satisfying $\sum_{i=1}^l m_i = n$. Then

$$\frac{1}{n} \sum_{j=1}^{\lfloor l/2 \rfloor} j(m_{l-2j+1} - m_{l-2j}) \leq 1/2 \quad (4)$$

with $m_0 = 0$

Proof: For simplicity of the notations, we assume l is odd, so that $l = 2a - 1$. The case l even follows in a similar way. Proving (4) by contradiction, we assume that

$$\sum_{j=1}^{a-1} j(m_{l-2j+1} - m_{l-2j}) > n/2. \quad (5)$$

Since

$$m_l + \sum_{j=1}^{a-1} (m_{l-2j+1} + m_{l-2j}) = n, \quad (6)$$

multiplying (5) by 2 and subtracting (6) from both sides yields, after shifting the pair grouping of consecutive values of m_i 's by one

$$\sum_{j=1}^{a-1} (2j - 1)(-m_{l-2j+2} + m_{l-2j+1}) - (2a - 1)m_1 > 0 \quad (7)$$

which is a contradiction as all the terms are nonpositive.

A variant-I PM code with $d_{\min}^2 = 2$ is optimum if and only if the three following conditions are satisfied (see, e.g., [2] for an equivalent formulation).

- Condition 1: As a set, $\{x_1, \dots, x_l\} = \{b, b + 1, \dots, b + l - 1\}$ for some real value b .
- Condition 2: $\sum_{i=1}^l m_i x_i = 0$.
- Condition 3: If $m_i \leq m_j$, then $|x_i| \geq |x_j|$.

Condition 1 ensures $d_{\min}^2 = 2$ while Condition 2 implies that although \mathbf{x} has dimension n , the effective dimension of the code is $n - 1$, so that the effective code rate becomes $R = \log_2 M / (n - 1)$.

In the following, we present a simple way to achieve these three conditions and for simplicity of the notations, we again assume l is

¹ $d_{\min}^2 = 2$ allows us to choose integer values for $x_i - x_{i-1}$'s [2].

odd; the results are extended in a straightforward way to the case when l is even. For $l = 2a - 1$ and $i = 1, \dots, l$, we assign to each m_i the value $a + k_i$, with $k_i = (-1)^{l-i} \lceil (l-i)/2 \rceil$. In other words, we assign a to m_l , $a - 1$ to m_{l-1} , $a + 1$ to m_{l-2} , \dots , 1 to m_2 , l to m_1 . It follows that

$$\frac{1}{n} \sum_{i=1}^l m_i (a + k_i) = a - \frac{1}{n} \sum_{j=1}^{a-1} j(m_{l-2j+1} - m_{l-2j}) \quad (8)$$

with $m_{l-2j+1} - m_{l-2j} \geq 0$.

Defining $x_i = a + k_i - (1/n) \sum_{i=1}^l m_i (a + k_i)$, we have

$$x_i = k_i + \frac{1}{n} \sum_{j=1}^{a-1} j(m_{l-2j+1} - m_{l-2j}). \quad (9)$$

As a result, Conditions 1 and 2 are straightforwardly satisfied. Condition 3 follows from Lemma 2.1, which ensures that the monotonicity of the values $|k_i|$ is preserved.

Note that this result also validates the optimum construction presented in [5] which implicitly assumes that the value a_m has the same number of terms to its left and its right.

Finally, as indicated in Section I, optimum variant II-a and variant II-b codes are obtained in a straightforward way with for $i = 1, \dots, l$, $x_i = l - i$ and $x_i = l - i + 1/\sqrt{2}$, respectively.

REFERENCES

- [1] D. Slepian, “Permutation modulation,” *Proc. IEEE*, vol. 53, no. 3, pp. 228–236, Mar. 1965.
- [2] E. Biglieri and M. Elia, “Optimum permutation modulation codes and their asymptotic performance,” *IEEE Trans. Inf. Theory*, vol. IT-23, no. 6, pp. 751–753, Nov. 1976.
- [3] D. Slepian, “Several new families of alphabets for signaling,” Bell Telephone Labs Unpublished Memorandum 1951.
- [4] I. Ingemarsson, “Optimized permutation modulation,” *IEEE Trans. Inf. Theory*, vol. 36, no. 5, pp. 1098–1100, Sep. 1990.
- [5] T. Ericson, INRIA, “Permutation Codes,” Rapport de Recherche INRIA 2109, Nov. 1993.